Filters.

Eugeniy E. Mikhailov

The College of William & Mary

Lecture 04

Notes

Power dissipation

Recall that power dissipated by element is

$$P = VI$$

where V and I are real.
Since we use a substitute
$V\cos(\omega t) \to Ve^{i\omega t}$ and $I\cos(\omega t) \to Ie^{i\omega t}$,
we need to write

$$P = Re(V)Re(I)$$

Recall the Ohm's law

$$V = ZI$$

Notes

Power dissipation by a reactive element

Theorem

Average power dissipated by a reactive element (C or L) is 0

Lets use as example an inductor.

$$Z_L = i\omega L = e^{i\frac{\pi}{2}}\omega L, I_L = I_p e^{i\omega t}$$

$$V_L = Z_L I_L = e^{i\frac{\pi}{2}}\omega L I_L = \omega L I_p e^{i(\omega t + \frac{\pi}{2})}$$

$$Re(I_L) = I_p\cos(\omega t), Re(V_L) = -\omega I_p L\sin(\omega t)$$

Thus average power dissipated by the inductor

$$P = \int_0^T Re(I_L)Re(V_L)dt = -\int_0^T I_p\cos(\omega t)\omega I_p L\sin(\omega t)dt$$

$$P = -\omega I_p^2 L\int_0^T \cos(\omega t)\sin(\omega t)dt = \omega I_p^2 L\int_0^T \frac{1}{2}\sin(2\omega t)dt = 0$$

Notes

Fourier transform

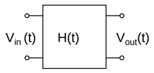
If function $f(t)$ goes to zero at $\pm\infty$ then $\hat{f}(w)$ exists such as

$$f(t) = \int_{-\infty}^{\infty}\hat{f}(\omega)e^{i\omega t}d\omega$$

Notes

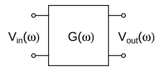
Transfer function

Time domain



$$V_{out}(t) = \int_{-\infty}^{t} H(t-\tau) V_{in}(\tau) d\tau$$

Frequency domain



$$V_{out}(\omega) = G(\omega) V_{in}(\omega)$$

Where G is complex transfer function or gain.

Definition

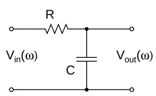
$$G(\omega) = \frac{V_{out}(\omega)}{V_{in}(\omega)} = |G(\omega)| e^{i\phi(\omega)}$$

Often used values of G in dB

$$dB = 20 \log_{10}(|G(w)|)$$

Notes

Simple example: RC low-pass filter



$$G(\omega) = \frac{V_{out}(\omega)}{V_{in}(\omega)} = \frac{\frac{1}{i\omega C}}{R + \frac{1}{i\omega C}} = \frac{1}{i\omega RC} \frac{1}{1 + \frac{1}{i\omega RC}} = \frac{1}{1 + i\omega RC}$$

defining $\omega_{3dB} = \frac{1}{RC}$

$$G(\omega) = \frac{1}{1 + i\frac{\omega}{\omega_{3dB}}} = \frac{1}{\sqrt{1 + \frac{\omega^2}{\omega_{3dB}^2}}} e^{i\phi}, \phi = atan(-\frac{\omega}{\omega_{3dB}})$$

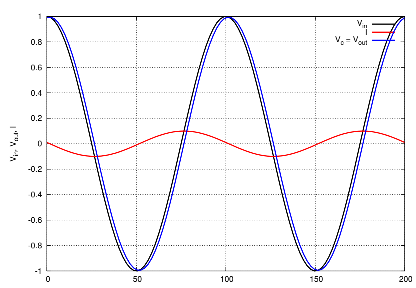
Note

$$|G(\omega = \omega_{3dB})| = 20 \log_{10}\left(\frac{1}{\sqrt{1+1}}\right) = 20 \log_{10}\left(\frac{1}{\sqrt{2}}\right) = -3dB$$

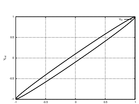
Notes

RC low-pass filter at $\omega = .1/RC$

Signal vs time



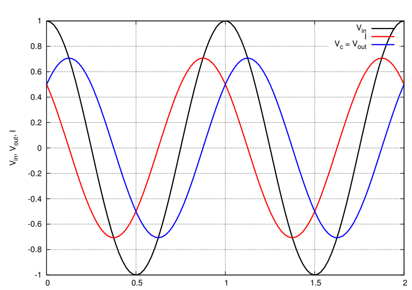
Lissajous plot



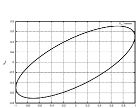
Notes

RC low-pass filter at $\omega = 1/RC$

Signal vs time



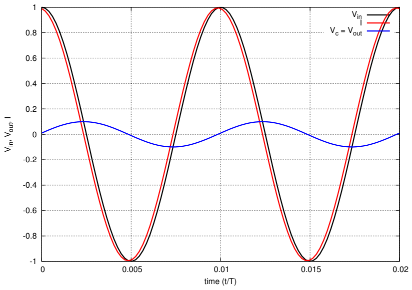
Lissajous plot



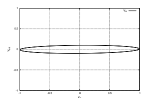
Notes

RC low-pass filter at $\omega = 10/RC$

Signal vs time



Lissajous plot

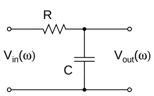


Notes

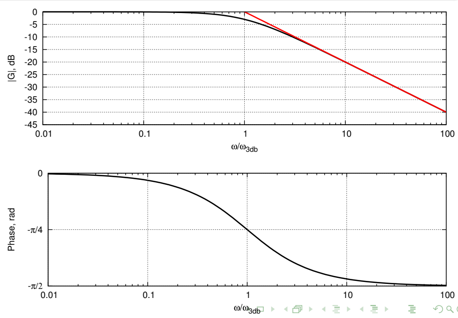
Bode plots

Definition

Bode plot: plots of magnitude and phase of the transfer function, where $|G|$ is often plotted in dB

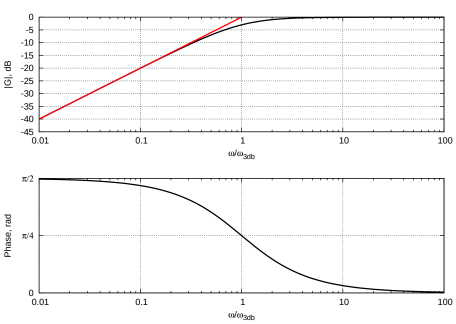


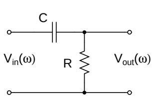
$$G(\omega) = \frac{1}{1 + i\frac{\omega}{\omega_{3dB}}}$$



Notes

RC high-pass filter





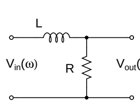
$$G(\omega) = \frac{V_{out}(\omega)}{V_{in}(\omega)} = \frac{R}{R + \frac{1}{i\omega C}} = \frac{i\omega RC}{1 + i\omega RC} = \frac{i\frac{\omega}{\omega_{3dB}}}{1 + i\frac{\omega}{\omega_{3dB}}}$$

with $\omega_{3dB} = \frac{1}{RC}$

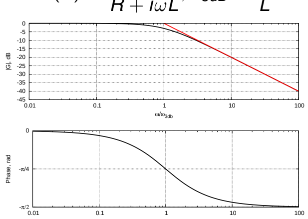
Notes

RL filters

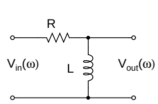
RL low-pass filter



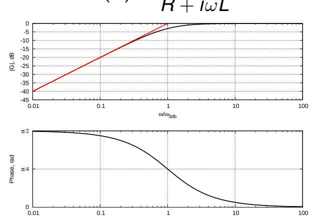
$$G(\omega) = \frac{R}{R + i\omega L}, \omega_{3dB} = \frac{R}{L}$$



RL high-pass filter

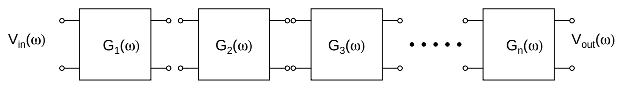


$$G(\omega) = \frac{i\omega L}{R + i\omega L}$$



Notes

Filters chain

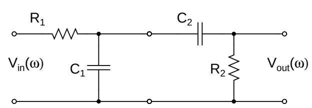


Technically next stage loads the previous and it is quite hard to calculate total transfer function.
However if we use rule of 10 to avoid overloading the previous filter. Every next stage resistor $R_{i+1} > 10R_i$ we can approximate

$$G_t(\omega) \approx G_1(\omega)G_2(\omega)G_3(\omega)\cdots G_n(\omega)$$

Notes

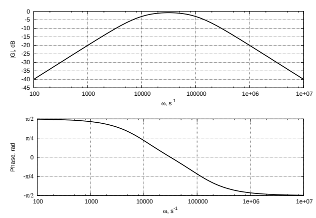
Example band pass filter



$$G_t(\omega) \approx G_1(\omega)G_2(\omega)$$

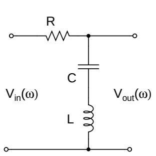
$$G_t(\omega) \approx \frac{1}{1+i\frac{\omega}{\omega_{1_{3dB}}}} \frac{i\frac{\omega}{\omega_{2_{3dB}}}}{1+i\frac{\omega}{\omega_{2_{3dB}}}}$$

For $R_1 = 1k\Omega$, $R_2 = 100k\Omega$, $C_1 = C_2 = .01\mu F$

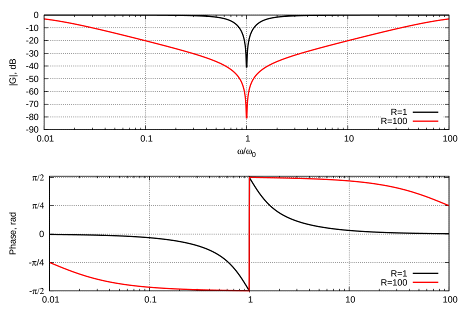


Notes

Notch filter - Band stop filter



$$\omega_0 = \frac{1}{\sqrt{LC}}$$



It is impossible to make band stop filter with $G \approx 1$ with only R,C or R,L elements. R, L, and C combo is required.

Notes

Notes